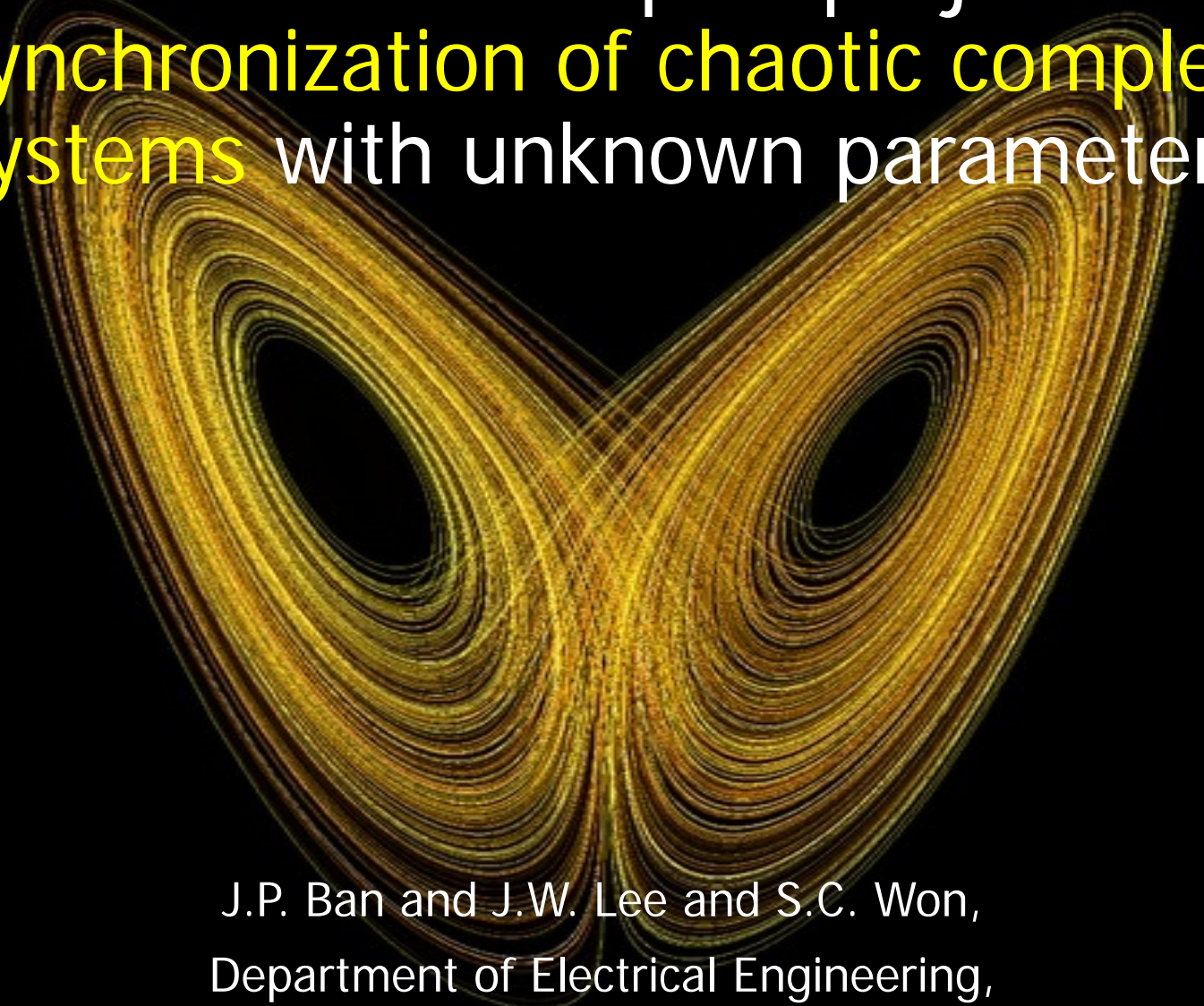


# Generalized complex projective synchronization of chaotic complex systems with unknown parameters



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# Presentation Outline

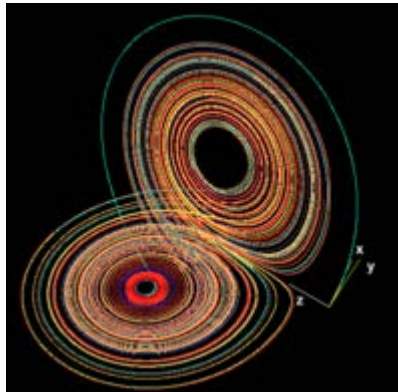
- Chaotic system
  - Synchronization
  - Complex number
- Problem Formulation
- Adaptive synchronization
  - Theorem
  - Proof of Stability
- Result
  - Numerical example
  - Simulation result
- Conclusion

# Small difference in **initial conditions** yield widely diverging outcomes

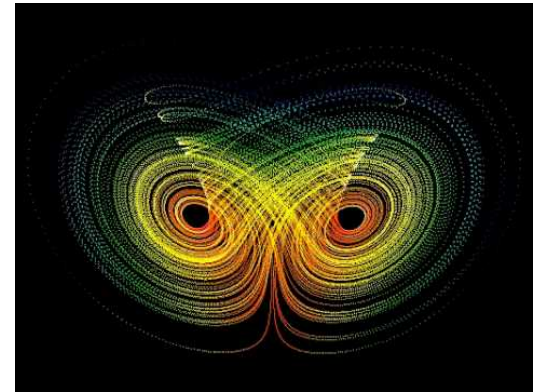
Examples of Chaotic system



Weather system



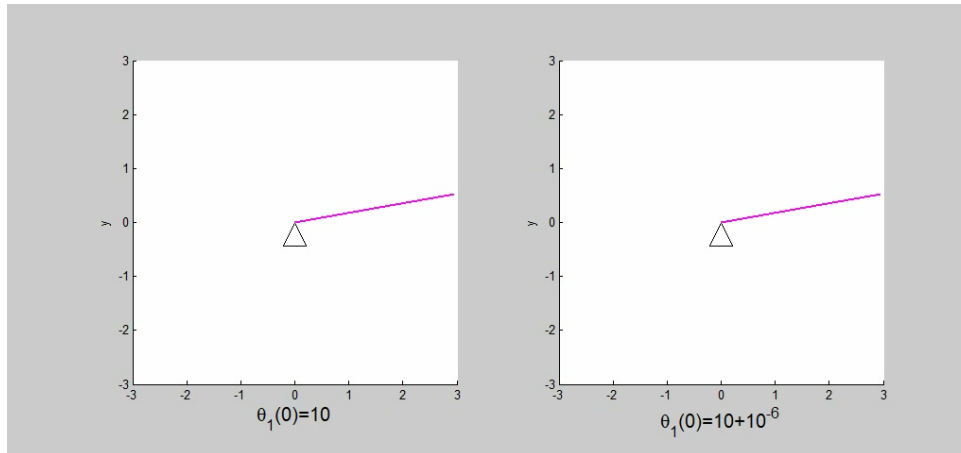
Lorenz attractor



Chen attractor

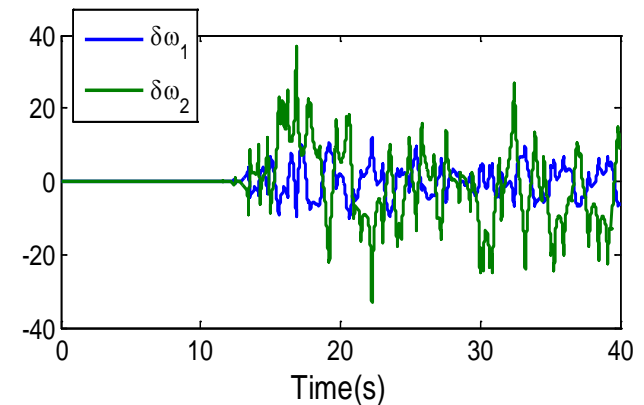
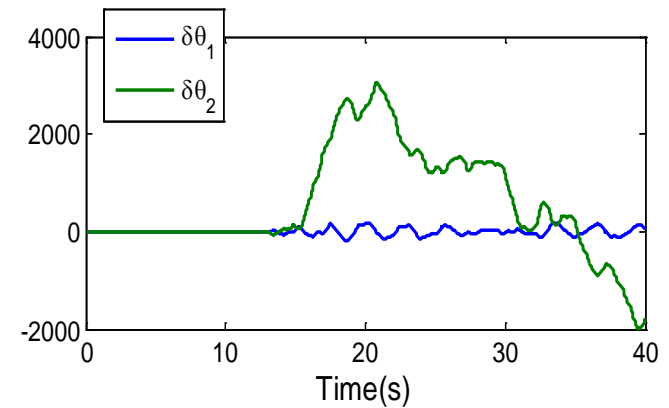
# Small difference in initial conditions yield widely diverging outcomes

Chaotic behavior of double rod pendulum



Initial conditions:  $\theta(0) = 10$

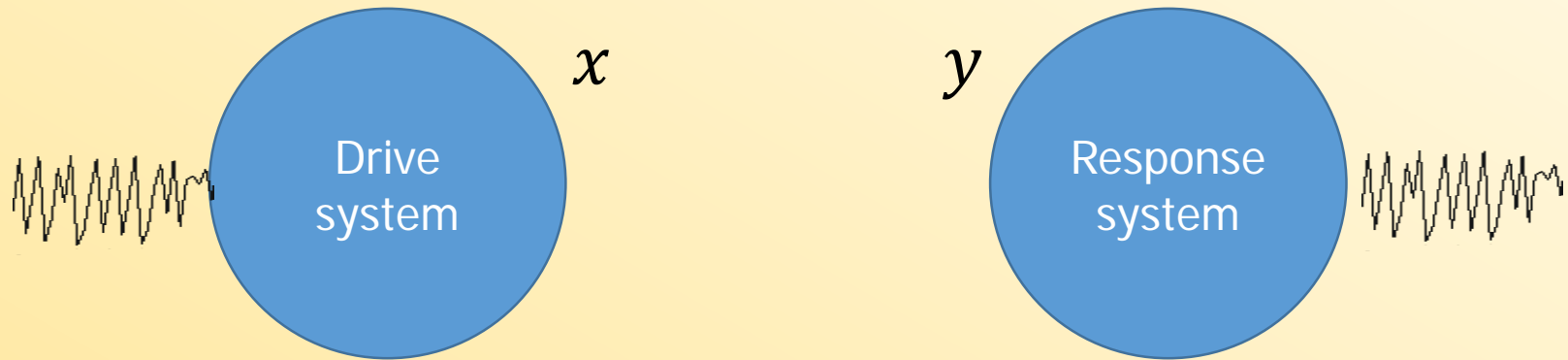
$\theta(0) = 10 + 10^{-6}$



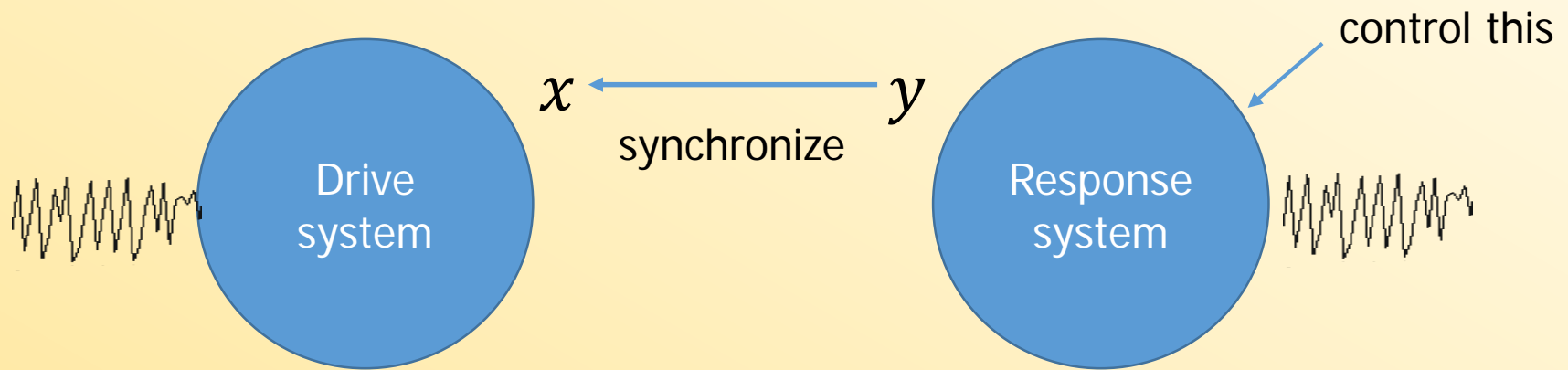
Synchronization is to make  
states of two systems same



Synchronization is to make  
**states** of two systems **same**

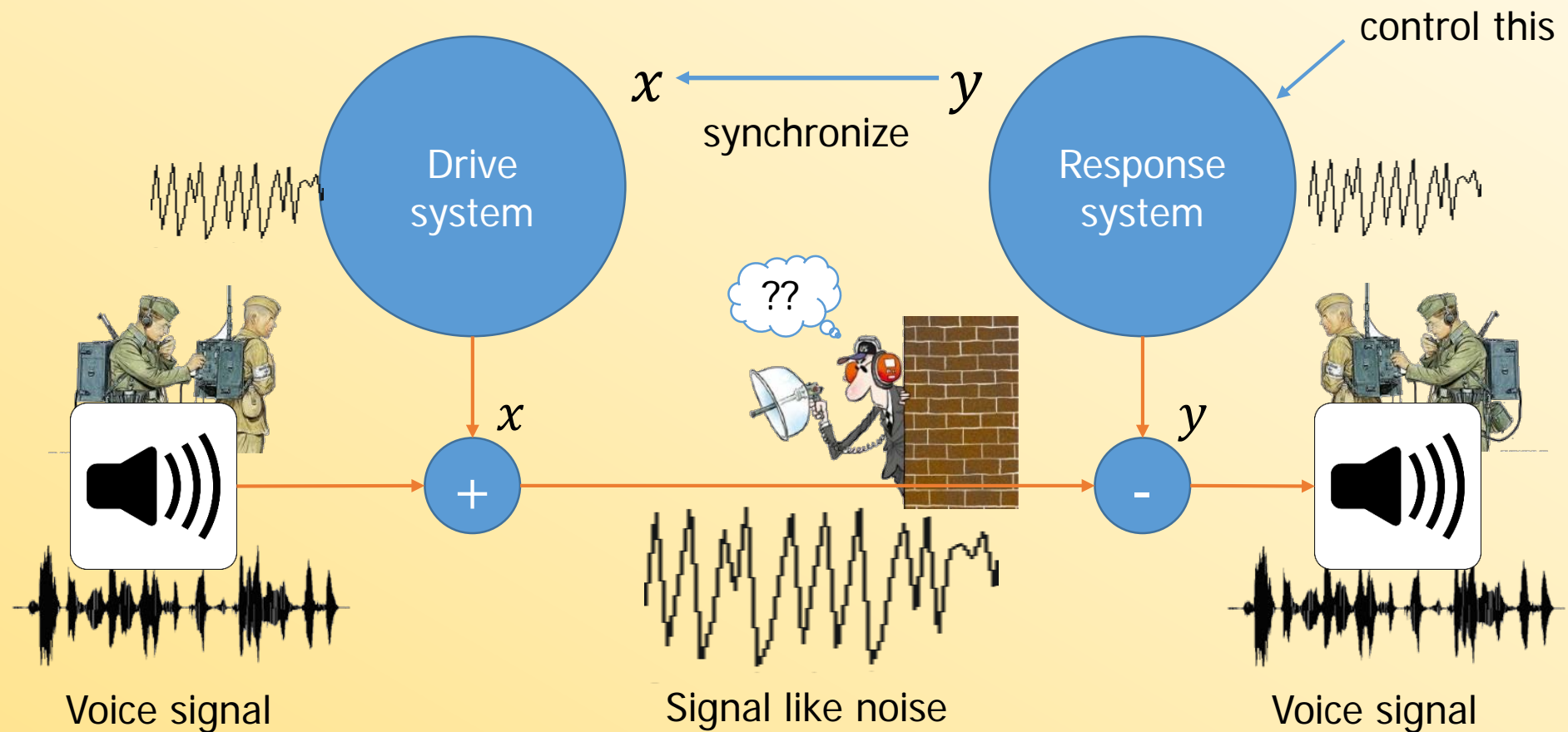


# Synchronization is to make states of two systems same





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# What we did is..

- Synchronization of chaotic system with:
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- Synchronization of chaotic system with:
  - Complex number states
  - Different dimensional
  - Unknown parameter

# Problem Formulation

- Consider the following drive-response chaotic complex systems

$$\dot{\mathbf{x}} = f(\mathbf{x}) + F(\mathbf{x})\theta \quad (1)$$

$$\dot{\mathbf{y}} = g(\mathbf{y}) + G(\mathbf{y})\varphi + \mathbf{u}(t), \quad (2)$$

$\mathbf{x}(t) = [x_1, \dots, x_m]^T$  is a state complex vector of the drive system (1),

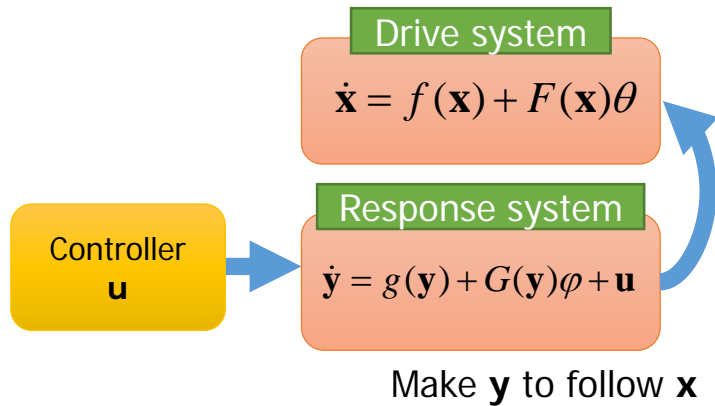
$\mathbf{y}(t) = [y_1, \dots, y_n]^T$  is a state complex vector of the response system (2),

- Define the error vector

$$\mathbf{e}(t) = \mathbf{y}(t) - H\mathbf{x}(t), \quad (3)$$

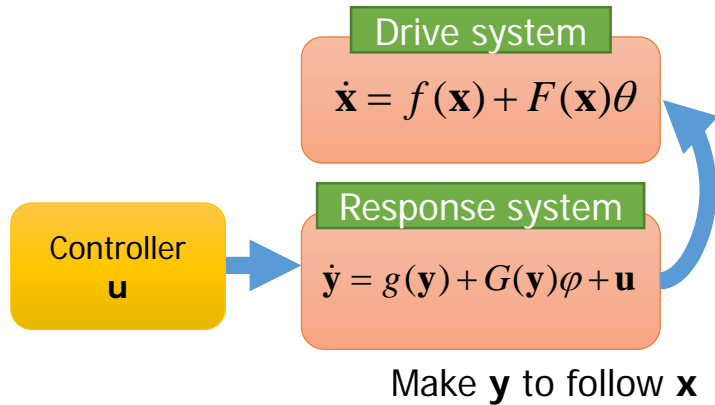
$H \in C^{n \times m}$  is a nonzero complex projective matrix whose components are constant complex numbers.

# Method



Subscript 'r' and 'i' represent real parts and imaginary parts of a complex matrix or vector or variable.

# Method

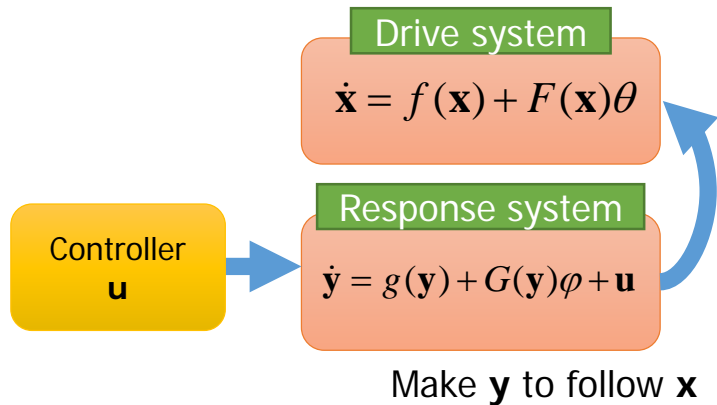


1. Use scaling matrix to match dimensions

$$e(t) = y(t) - Hx(t),$$

Subscript 'r' and 'i' represent real parts and imaginary parts of a complex matrix or vector or variable.

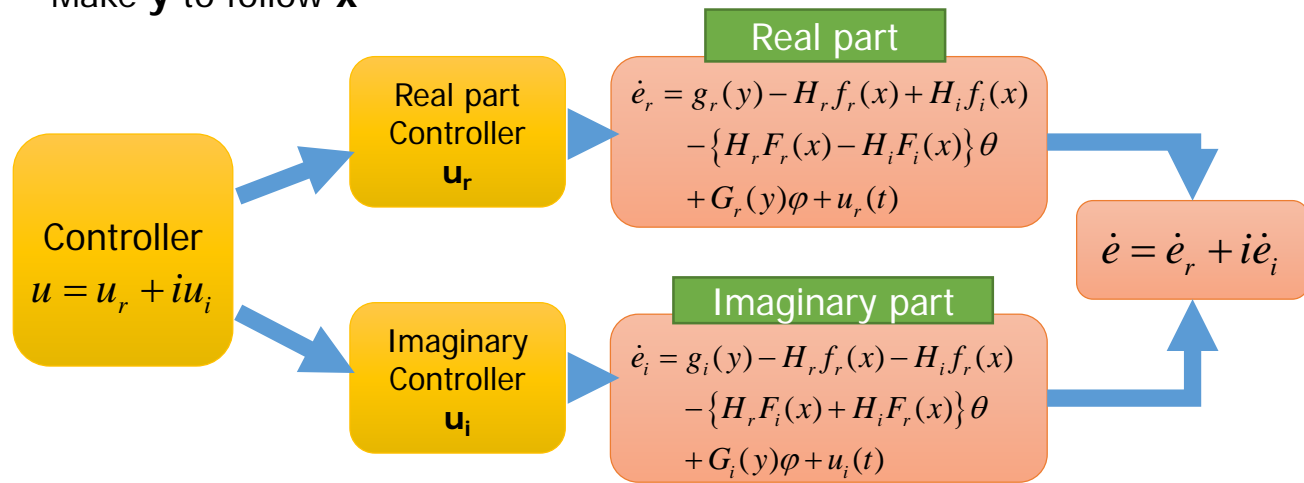
# Method



1. Use scaling matrix to match dimensions

$$e(t) = y(t) - Hx(t),$$

2. Derive error dynamics separately & Design controllers separately into real and imaginary



Subscript 'r' and 'i' represent real parts and imaginary parts of a complex matrix or vector or variable.



# Control inputs and Update laws for an adaptive controller

Theorem 1: *For the chaotic complex drive system (1), the response system (2) and the scaling matrix  $H$ , the Adaptive generalized complex projective synchronization (AGCPS) of the chaotic complex systems is accomplished if a control input and adaptive laws are chosen as follows*

- Control inputs

$$\begin{aligned} u_r(t) = & -g_r(y) + H_r F_r(x) - H_i F_i(x) \\ & + \{H_r F_r(x) - H_i F_i(x)\} \hat{\theta} \\ & - G_r(y) \hat{\phi} - K e_r(t) \end{aligned} \quad (4)$$

$$\begin{aligned} u_i(t) = & -g_i(y) + H_r F_i(x) + H_i F_r(x) \\ & + \{H_r F_i(x) + H_i F_r(x)\} \hat{\theta} \\ & - G_i(y) \hat{\phi} - K e_i(t) \end{aligned} \quad (5)$$

- Update Laws

$$\begin{aligned} \dot{\hat{\theta}} = & -K_\theta \left[ \{H_r F_r(x) - H_i F_i(x)\}^T e_r(t) \right. \\ & \left. + \{H_r F_i(x) + H_i F_r(x)\}^T e_i(t) \right] \end{aligned} \quad (6)$$

$$\dot{\hat{\phi}} = K_\phi \left[ G_r(y)^T e_r(t) + G_i(y)^T e_i(t) \right] \quad (7)$$

'hat' represents the estimation of parameter vector of a complex chaotic systems.

# Proof of the theorem

- Choose the following Lyapunov function candidate:

$$V(t) = \frac{1}{2} \{ \mathbf{e}_r^T \mathbf{e}_r + \mathbf{e}_i^T \mathbf{e}_i \} + \frac{1}{2} K_\theta^{-1} \tilde{\boldsymbol{\theta}}^T \tilde{\boldsymbol{\theta}} + \frac{1}{2} K_\phi^{-1} \tilde{\boldsymbol{\phi}}^T \tilde{\boldsymbol{\phi}}, \quad (8)$$

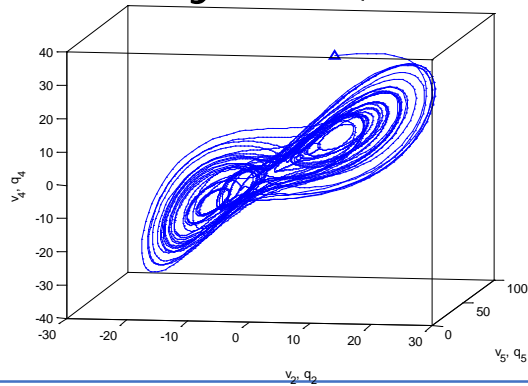
where  $\tilde{\boldsymbol{\theta}} = \hat{\boldsymbol{\theta}} - \boldsymbol{\theta}$ , and  $\tilde{\boldsymbol{\phi}} = \hat{\boldsymbol{\phi}} - \boldsymbol{\phi}$ .

- Take a time derivative of (8) and substituting (4)~(7) yields

$$\begin{aligned} \dot{V}(t) &= \mathbf{e}_r^T \left[ \{ H_r F_r(\mathbf{x}) - H_i F_i(\mathbf{x}) \} \tilde{\boldsymbol{\theta}} - G_r(\mathbf{y}) \tilde{\boldsymbol{\phi}} - K \mathbf{e}_r(t) \right] \\ &\quad + \mathbf{e}_i^T \left[ -\{ H_r F_i(\mathbf{x}) + H_i F_r(\mathbf{x}) \} \tilde{\boldsymbol{\theta}} - G_i(\mathbf{y}) \tilde{\boldsymbol{\phi}} - K \mathbf{e}_i(t) \right] \\ &\quad + K_\theta^{-1} \tilde{\boldsymbol{\theta}}^T \dot{\hat{\boldsymbol{\theta}}} + K_\phi^{-1} \tilde{\boldsymbol{\phi}}^T \dot{\hat{\boldsymbol{\phi}}}. \\ &= -\mathbf{e}_r^T K \mathbf{e}_r(t) - \mathbf{e}_i^T K \mathbf{e}_i(t) \\ &= -\sum_{j=1}^n k_j \left[ e_{rj}^2 + e_{ij}^2 \right] \leq k_{\min} \|\mathbf{e}(t)\| \end{aligned}$$

# Numerical Simulations

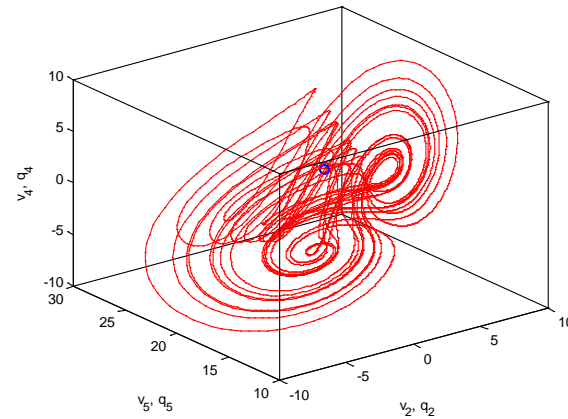
- Hyperchaotic complex Lorenz-type system (Drive system)



$$\begin{cases} \dot{x}_1 = \alpha(x_2 - x_1) \\ \dot{x}_2 = \gamma x_1 - x_2 - x_1 x_3 + x_4 \\ \dot{x}_3 = \frac{1}{2}(\bar{x}_1 x_2 + x_1 \bar{x}_2) - \beta x_3 \\ \dot{x}_4 = \rho x_1 + \mu x_2 \end{cases}$$

$$\alpha = 14, \beta = 3, \gamma = 45, \rho = -5, \mu = -4$$

- Complex Chen system (Response System)



$$\begin{cases} \dot{y}_1 = a(y_2 - y_1) + u_1 \\ \dot{y}_2 = (b - a)y_1 + by_2 - y_1 y_3 + u_2 \\ \dot{y}_3 = \frac{1}{2}(\bar{y}_1 y_2 + y_1 \bar{y}_2) - cy_3 + u_3 \end{cases}$$

$$a = 28, b = 22, c = 1$$

# Numerical Simulations

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- Complex projective matrix

$$H = \begin{bmatrix} 0.1 + 0.5i & 0.1i & 0 & 0 \\ 0 & 0.5 + 0.2i & 0.5 + 0.1i & 0 \\ 0 & 0 & 0.1 & 0.2 + 0.5i \end{bmatrix}$$

# Numerical Simulations

- We can represent the drive system and the response system as a form of (1) and (2), then

- Drive system

$$\dot{\mathbf{x}} = f_r(\mathbf{x}) + F_r(\mathbf{x})\theta + i\{f_i(\mathbf{x}) + F_i(\mathbf{x})\theta\},$$

$$f_r(\mathbf{x}) = \begin{bmatrix} 0 \\ -x_{r1}x_{r3} - x_{r3} + x_{r4} \\ x_{r1}x_{r2} + x_{i1}x_{i2} \\ 0 \end{bmatrix}, \quad f_i(\mathbf{x}) = \begin{bmatrix} 0 \\ -x_{i1}x_{r3} - x_{i2} + x_{i4} \\ 0 \\ 0 \end{bmatrix},$$

$$F_r(\mathbf{x}) = \begin{bmatrix} x_{r2} - x_{r1} & 0 & 0 & 0 & 0 \\ 0 & 0 & x_{r1} & 0 & 0 \\ 0 & -x_{r3} & 0 & 0 & 0 \\ 0 & 0 & 0 & x_{r1} & x_{r2} \end{bmatrix},$$

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$$\theta = [\alpha \quad \beta \quad \gamma \quad \rho \quad \mu]^T$$

- Response system

$$\dot{\mathbf{y}} = g_r(\mathbf{y}) + G_r(\mathbf{y})\phi + i\{g_i(\mathbf{y}) + G_i(\mathbf{y})\phi\} + \mathbf{u}(t),$$

$$g_r(\mathbf{x}) = \begin{bmatrix} 0 \\ -y_{r1}y_{r3} \\ y_{r1}y_{r3} + y_{i1}y_{i2} \end{bmatrix}, \quad g_i(\mathbf{x}) = \begin{bmatrix} 0 \\ -y_{i1}y_{i3} \\ 0 \end{bmatrix},$$

$$G_r(\mathbf{x}) = \begin{bmatrix} y_{r2} - y_{r1} & 0 & 0 \\ -y_{r1} & y_{r1} + y_{r2} & 0 \\ 0 & 0 & -y_{r3} \end{bmatrix},$$

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$$\phi = [a \quad b \quad c]^T$$

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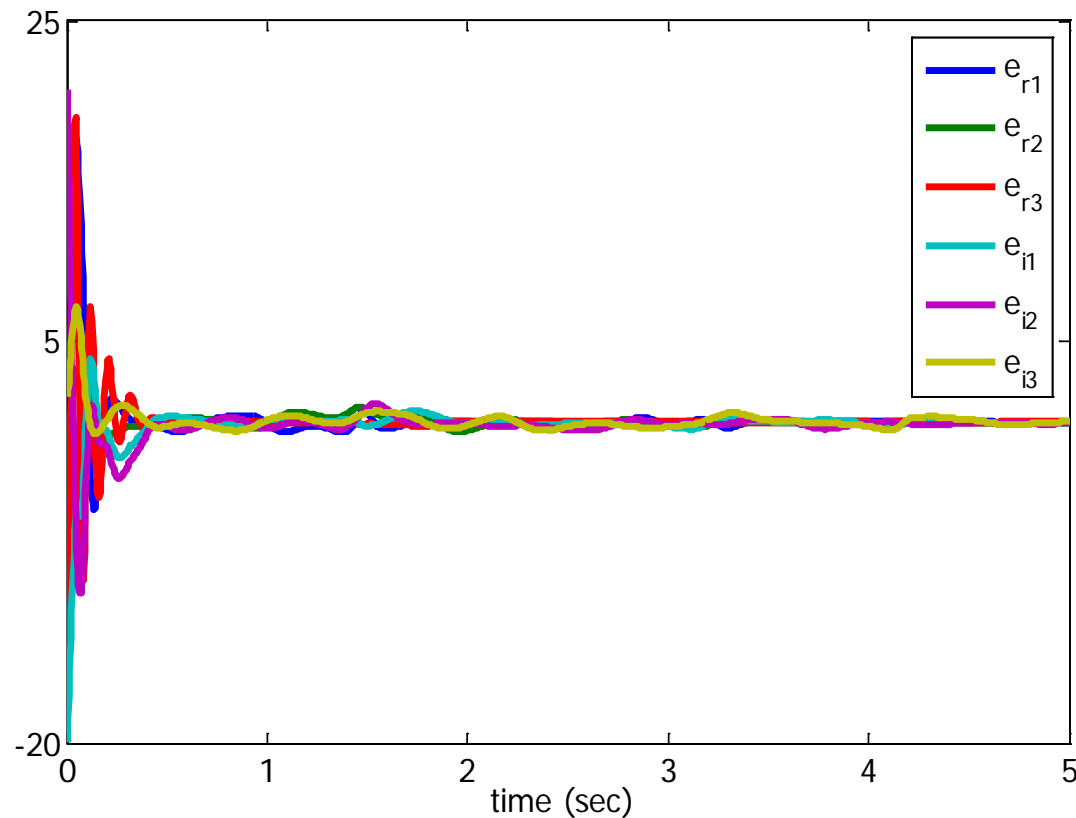
$$\begin{aligned} \dot{\hat{\theta}} = & -K_\theta \left[ \{H_r F_r(x) - H_i F_i(x)\}^T e_r(t) \right. \\ & \left. + \{H_r F_i(x) + H_i F_r(x)\}^T e_i(t) \right] \end{aligned} \quad (6)$$

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# States are synchronized by using adaptive controller

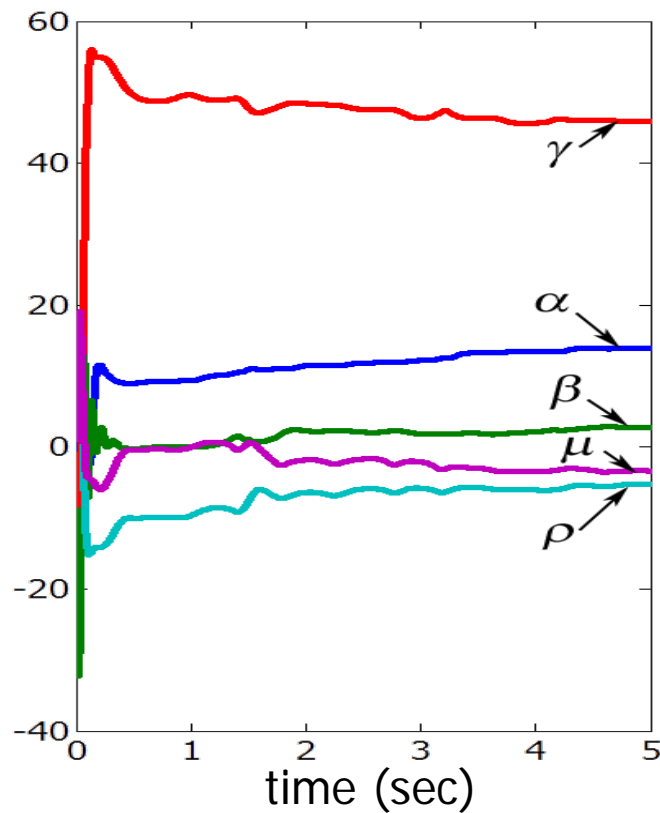
- Synchronization errors in an example



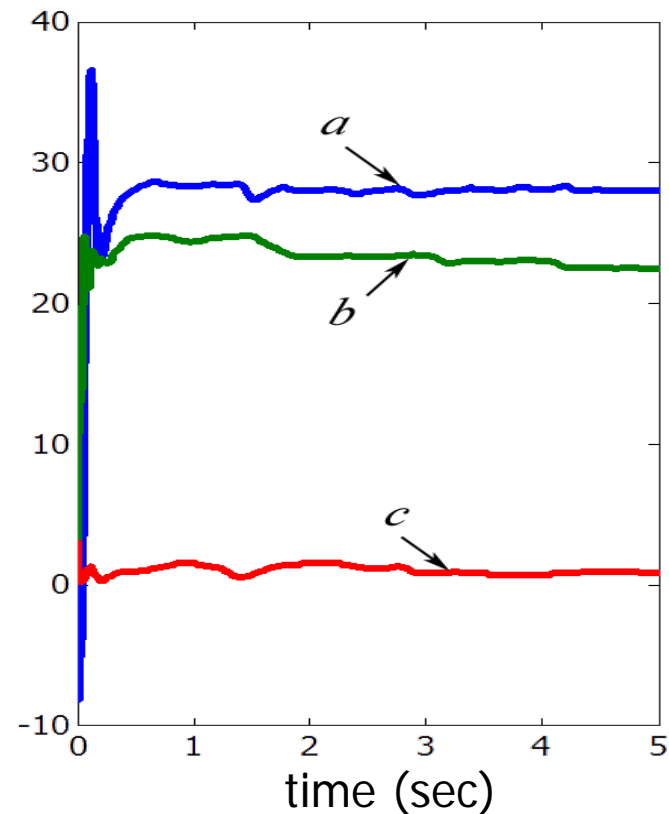


# Parameters converge to actual values using adaptive law

< Parameters of Drive system >



< Parameters of Response system >



# Conclusions:

- With complex number, chaotic systems becomes much more complicated.
- Complex number chaotic systems can be synchronized although the parameters of the systems are unknown.
- Using scaling matrix, synchronizations of chaotic systems which have different dimension can be accomplished.